# Robust Exponential Stability for Uncertain Discrete-Time Switched Systems with Interval Time-Varying Delay through a Switching Signal

Jeng-Der Chen<sup>1\*</sup>, I-Te Wu<sup>2</sup>, Chang-Hua Lien<sup>3</sup>, Chin-Tan Lee<sup>4</sup>, Ruey-Shin Chen<sup>5</sup> and Ker-Wei Yu<sup>6</sup>

<sup>1,5</sup> Department of Computer Science and Information Engineering National Quemoy University, Kinmen,

Taiwan, R.O.C.

\*tdchen@ngu.edu.tw.

<sup>2</sup> Department of Life Science-National Taiwan Normal University and Department of Sports and Leisure National Quemoy University Kinmen, Taiwan, R.O.C.

Kinmen, Taiwan, R.O.C.

3.6 Department of Marine Engineering, National Kaohsiung Marine University
Kaohsiung, Taiwan, R.O.C.

Kaohsiung, Taiwan , R.O.C.

Department of Electronic Engineering, National Quemoy University Kinmen, Taiwan , R.O.C.

#### **ABSTRACT**

This paper deals with the switching signal design to robust exponential stability for uncertain discrete-time switched systems with interval time-varying delay. The lower and upper bounds of the time-varying delay are assumed to be known. By construction of a new Lyapunov-Krasovskii functional and employing linear matrix inequality, some novel sufficient conditions are proposed to guarantee the global exponential stability for such system with parametric perturbations by using a switching signal. In addition, some nonnegative inequalities are used to provide additional degrees of freedom and reduce the conservativeness of systems. Finally, some numerical examples are given to illustrate performance of the proposed design methods.

Keywords: Switching signal, exponential stability, discrete switched systems, interval time-varying delay, additional nonnegative inequality.

#### 1. Introduction

A switched system is composed of a family of subsystems and a switching signal that specifies which subsystem is activated to the system trajectories at each instant of time [1]. Some real examples for switched systems are automated highway systems, automotive engine control system, chemical process, constrained robotics, power systems and power electronics, robot manufacture, and stepper motors [2-14]. Switching among linear systems may produce many complicate system behaviors, such as multiple limit cycles and chaos [1]. It is also well known that the existence of delay in a system may cause instability or bad performance in closed control systems [15-17]. Time-delay phenomena are usually appeared in many practical systems, such as AIDS epidemic, chemical engineering systems, hydraulic systems, inferred grinding model, neural network, nuclear reactor, population dynamic model, and rolling mill. Hence stability analysis and

stabilization for discrete switched systems with time delay have been researched in recent years [2-4, 7, 9-10, 12-14].

It is interesting to note that the stable property for each subsystem cannot imply that the overall system is also stable under arbitrary switching signal [4, 6, 8-9]. Another interesting fact is that the stability of a switched system can be achieved by choosing the switching signal even when each subsystem is unstable [5, 7-8]. In [7], a switching signal design technique is proposed to guarantee the asymptotic stability of discrete switched systems with interval time-varying delay. In [14], the switching signal is identified to guarantee the stability of discrete switched time-delay system. Additional nonnegative inequality terms had been used to improve the conservativeness for the obtained results in our past results [4-5]. Hence the global exponential stability problems for uncertain switched discrete systems with interval timevarying delay under some certain switching signal will be considered in this paper. Some numerical examples are provided to demonstrate the use of obtained results. From the simulation results, our proposed approach provides some less conservative results.

The notations are used throughout this paper. For a matrix A, we denote the transpose by  $A^T$ , spectral norm by  $\|A\|$ , symmetric positive (negative) definite by A>0 (A<0), maximal eigenvalue by  $\lambda_{\max}(A)$ , minimal eigenvalue by  $\lambda_{\min}(A)$ ,  $n\times m$  dimension by  $A_{(n\times m)}$ .  $A\leq B$  means that matrix B-A is symmetric positive semi-definite. 0 and I denote the zero matrix and identity matrix with compatible dimensions.  $diag\{\cdots\}$  stands for a block-diagonal matrix. For a vector x, we denote the Euclidean norm by  $\|x\|$ . Define  $\overline{N}=\{1,2\cdots,N\}$ , N>1,  $\|x_k\|_s=\max_{\theta=r_k-r_k+r_k+1\cdots 0}\|x(k+\theta)\|$ .

## 2. Problem statement and preliminaries

Consider the following uncertain discrete switched system with interval time-varying delay of the form

$$x(k+1) = \overline{A}_{\sigma(k)}x(k) + \overline{A}_{d\sigma(k)}x(k-r(k))$$

$$x(\theta) = \phi(\theta), \ \theta = -r_M, -r_M + 1, \dots, 0,$$
 (1)

where  $x(k) \in \Re^n$  is the state vector,  $x_k$  is the state at time k defined by  $x_k(\theta) := x(k+\theta), \ \forall \theta \in \{-r_M, -r_M+1, \cdots, 0\}$ , switching signal law  $\sigma(k, x(k))$  is a piecewise constant function depending on the state in each time. The switching signal  $\sigma(k, x(k))$  takes its value in the finite set  $\overline{N}$  and will be designed. Moreover, the  $\sigma(k) = i$  implied that the i-th subsystem  $(\overline{A}_i, \overline{A}_{di})$  is activated, and  $\phi(k) \in \Re^n$  is a vector-valued initial state function. The time-varying delay r(k) is a function from  $\{0, 1, 2, 3, \cdots\}$  to  $\{0, 1, 2, 3, \cdots\}$ , and satisfies the following condition:

$$0 < r_m \le r(k) \le r_M \tag{2}$$

where  $r_{\scriptscriptstyle m}$  and  $r_{\scriptscriptstyle M}$  are two given positive integers.

The matrices  $\overline{A}_i, \overline{A}_{di}$  are assumed to be  $\overline{A}_i = A_i + \Delta A_i(k)$ ,  $\overline{A}_{di} = A_{di} + \Delta A_{di}(k)$ , where  $A_i, A_{di} \in \Re^{n \times n}$ ,  $i = 1, 2, \cdots, N$ , are some given constant matrices.  $\Delta A_i(k)$ ,  $\Delta A_{di}(k)$ ,  $i = 1, 2, \cdots, N$ , are some perturbed matrices and satisfy the following conditions

$$\left[\Delta A_{i}(k) \quad \Delta A_{di}(k)\right] = M_{i}F_{i}(k)\left[N_{1i} \quad N_{2i}\right] \tag{3}$$

where  $M_i$  and  $N_{ji}$ , j=1,2  $i=1,2,\cdots,N$ , are some given constant matrices of system with appropriate dimensions, and  $F_i(k)$ ,  $i=1,2,\cdots,N$ , are unknown matrices representing the parameter perturbations which satisfy

$$F_i^T(k)F_i(k) \le I \tag{4}$$

In order to derive our main results, the Lemmas are introduced as follows:

**Lemma 1.** [4] Let U, V, W and M be real matrices of appropriate dimensions with M satisfying  $M = M^T$ , then  $M + UVW + W^TV^TU^T < 0$  for all  $V^TV \leq I$ , if and only if there exists a scalar  $\varepsilon > 0$  such that  $M + \varepsilon^{-1} \cdot UU^T + \varepsilon \cdot W^TW = M + \varepsilon^{-1} \cdot UU^T + \varepsilon^{-1} \cdot \left(\varepsilon W\right)^T \left(\varepsilon W\right) < 0$ 

**Lemma 2.** (Schur complement of [18]). For a given matrix  $W = \begin{bmatrix} W_{11} & W_{12} \\ W_{12}^T & W_{22} \end{bmatrix}$  with  $W_{11} = W_{11}^T$ ,  $W_{22} = W_{22}^T$ ,

then the following conditions are equivalent:

(1) 
$$W < 0$$
, (2)  $W_{22} < 0$ ,  $W_{11} - W_{12}W_{22}^{-1}W_{12}^{T} < 0$ .

Define the some sets

$$\Omega_{i}(\alpha, P, U, A_{i}) = \left\{ x \in \Re^{n} : x^{T}(\alpha^{-2} \cdot A_{i}^{T}PA_{i} - P + U)x \leq 0 \right\},$$
where  $0 < \alpha < 1$ ,  $P > 0$ , and  $Q > 0$ , then we obtain

$$\overline{\Omega}_{1} = \Omega_{1}, \ \overline{\Omega}_{2} = \Omega_{1} - \overline{\Omega}_{1}, \cdots, 
\overline{\Omega}_{i} = \Omega_{i} - \bigcup_{i=1}^{i-1} \overline{\Omega}_{j}, \cdots, \ \overline{\Omega}_{N} = \Omega_{N} - \bigcup_{i=1}^{N-1} \overline{\Omega}_{j}$$
(6)

From the above sets definition, then the switching law is designed as follows:

$$\sigma = i$$
, if  $x \in \overline{\Omega}_i$ ,  $i \in \overline{N}$ , (7)

where  $\overline{\Omega}_i$  is defined in (6).

The key lemma can be easily proved by the similar way in [7].

**Lemma 3.** If there exist some scalars  $0 \le \alpha \le 1$ ,  $0 \le \alpha_i \le 1$ ,  $i \in \overline{N}$ , and  $\sum_{i=1}^N \alpha_i = 1$ , some matrices

$$P > 0$$
 and  $Q > 0$ , such that  $\sum_{i=1}^{N} \alpha_i \cdot \begin{bmatrix} -P + U & A_i^T P \\ * & -\alpha^2 \cdot P \end{bmatrix} < 0$ 

Obviously, we can obtain  $\bigcup\limits_{i=1}^{N}\overline{\Omega}_{i}=\mathfrak{R}^{n}$  and  $\overline{\Omega}_{i}\cap\overline{\Omega}_{j}=\varphi$ ,  $i\neq j$ ,  $i,j=1,2,\cdots,N$ , where  $\varphi$  is an empty set and  $\overline{\Omega}_{i}$  is defined in (6).

**Remark 1.** Consider a discrete switched system:  $x(k+1) = A_{\sigma(k)}x(k)$ .

Now we can choose the Lyapunov function as  $V(x_k) = \alpha^{-2k} \cdot x^T(k) Px(k)$ 

where matrix P>0 . We can obtain  $\Delta V(x_k)=\alpha^{-2k}\cdot\left[\alpha^{-2}\cdot x^T(k+1)Px(k+1)-x^T(k)Px(k)\right].$ 

Assume  $\sigma(k, x(k)) = q \in \overline{N}$ , then we have  $\Delta V(x_k) = \alpha^{-2k} \cdot x^T(k) (\alpha^{-2} \cdot A_q^T P A_q - P + U - U) x(k)$ 

where the matrix U > 0.

If the condition in Lemma 3 is satisfied, then we can get

$$\begin{bmatrix} -P + U & A_q^T P \\ * & -\alpha^2 \cdot P \end{bmatrix} < 0$$
 (8)

By Lemma 2 with (8), then we can have the following result:

$$x^{T}(k)(\alpha^{-2} \cdot A_{q}^{T}PA_{q} - P + U)x(k) < 0, \quad \forall x(k) \in \overline{\Omega}_{q}$$
  
$$\Delta V(x_{k}) < 0$$

To achieve the exponential stability of discrete switched system, the condition in Lemma 3 is a reasonable choice and feasible setting.

#### 3. Main results

**Theorem 1.** System (1) is globally exponentially stable with convergence rate  $0<\alpha<1$  for any timevarying delay r(k) satisfying (2), if there exist some  $n\times n$  matrices P>0, S>0, T>0, Q>0, U>0,  $R_i>0$ , i=1,2,3,4,  $R_{j11}\in\Re^{4n\times 4n}$ ,  $R_{j12}\in\Re^{4n\times n}$ , j=1,2,3,4,5, and the positive numbers  $\mathcal{E}_q$ ,  $q=1,2,\cdots,N$ , such that the following LMI conditions are feasible:

$$\begin{bmatrix} R_{i11} & R_{i12} \\ * & R_i \end{bmatrix} > 0, \ i = 1, 2, 3, 4, 5, \ (R_4 = R_5), \tag{9}$$

$$\Xi_{jq} = \begin{bmatrix} \Xi_{11q}^{j} & \Xi_{12q} & \Xi_{13q} \\ * & \Xi_{22q} & \Xi_{23q} \\ * & * & \Xi_{33q} \end{bmatrix} < 0, j = 1, 2, q = 1, 2, \dots; N,$$
(10)

$$\sum_{i=1}^{N} \alpha_{i} \cdot \begin{bmatrix} -P + U & A_{i}^{T} P \\ * & -\alpha^{2} \cdot P \end{bmatrix} < 0$$
 (11)

where

$$\begin{split} \Xi_{11q}^{1} &= diag \left\{ -\alpha^{-2} \cdot A_{q}^{T} P A_{q} - U + S + \left(r_{Mm} + 1\right) \cdot Q + T \right. \\ &- \alpha^{2r_{m}} \cdot S - \alpha^{2r_{M}} \cdot Q - \alpha^{2r_{M}} \cdot T \right\} \\ &+ \alpha^{2r_{m}} \cdot \left(r_{m} \cdot R_{111} + R_{112} W_{1} + W_{1}^{T} R_{112}^{T}\right) \\ &+ \alpha^{2r_{M}} \cdot \left[r_{M} \cdot R_{211} + r_{M} \cdot R_{311} + r_{Mm} \cdot R_{511} \right. \\ &+ \sum_{j=2}^{5} \left(R_{j12} W_{j} + W_{j}^{T} R_{j12}^{T}\right), \\ \Xi_{11q}^{2} &= diag \left\{ -\alpha^{-2} \cdot A_{q}^{T} P A_{q} - U + S + \left(r_{Mm} + 1\right) \cdot Q + T \right. \\ &- \alpha^{2r_{m}} \cdot S - \alpha^{2r_{M}} \cdot Q - \alpha^{2r_{M}} \cdot T \right\} \end{split}$$

$$+\alpha^{2r_{m}} \cdot \left(r_{m} \cdot R_{111} + R_{112}W_{1} + W_{1}^{T}R_{112}^{T}\right) +\alpha^{2r_{M}} \cdot \left[r_{M} \cdot R_{211} + r_{M} \cdot R_{311} + r_{Mm} \cdot R_{411}\right]$$

$$+\sum_{i=2}^{5} \left( R_{j12} W_{j} + W_{j}^{T} R_{j12}^{T} \right)$$

$$\Xi_{12q} = \begin{bmatrix} A_q^T P & (A_q - I)^T R_1 \\ 0 & 0 \\ A_{dq}^T P & A_{dq}^T R_1 \\ 0 & 0 \end{bmatrix}$$

$$\begin{pmatrix} (A_q - I)^T R_2 & (A_q - I)^T R_3 & (A_q - I)^T R_4 \\ 0 & 0 & 0 \\ A_{dq}^T R_2 & A_{dq}^T R_3 & A_{dq}^T R_4 \\ 0 & 0 & 0 \end{pmatrix}$$

$$\Xi_{22q} = -diag\left\{\alpha^2 \cdot P \quad \frac{R_1}{r_m} \quad \frac{R_2}{\hat{r}_{Mm}} \quad \frac{R_3}{r_M} \quad \frac{R_4}{r_{Mm}}\right\}$$

$$\Xi_{13q} = \begin{bmatrix} 0 & \varepsilon_q \cdot N_{1q}^T \\ 0 & 0 \\ 0 & \varepsilon_q \cdot N_{2q}^T \\ 0 & 0 \end{bmatrix}, \ \Xi_{23q} = \begin{bmatrix} PM_q & 0 \\ R_1M_q & 0 \\ R_2M_q & 0 \\ R_3M_q & 0 \\ R_4M_q & 0 \end{bmatrix},$$

$$\Xi_{33q} = -diag\{\varepsilon_q \cdot I \quad \varepsilon_q \cdot I\}, \ r_{Mm} = r_M - r_m,$$

$$\hat{r}_{Mm} = \frac{1}{2} (r_{Mm} + 1) \cdot (r_M + r_m), W_1 = \begin{bmatrix} I & -I & 0 & 0 \end{bmatrix}_{(n \times 4n)},$$

$$W_2 = \begin{bmatrix} I & 0 & -I & 0 \end{bmatrix}_{(n \times 4n)}, W_3 = \begin{bmatrix} I & 0 & 0 & -I \end{bmatrix}_{(n \times 4n)},$$

$$W_4 = \begin{bmatrix} 0 & 0 & I & -I \end{bmatrix}_{(n \times 4n)}, \ W_5 = \begin{bmatrix} 0 & I & -I & 0 \end{bmatrix}_{(n \times 4n)}.$$

**Proof.** The Lyapunov-Krasovskii functionals candidate is given by

$$V(x_k) = \sum_{i=1}^{8} V_i(x_k)$$
 (12)

where

$$V_1(x_k) = \alpha^{-2k} \cdot x^T(k) Px(k)$$
(13)

$$V_2(x_k) = \sum_{\theta = k - r_m}^{k - 1} \alpha^{-2\theta} \cdot x^T(\theta) Sx(\theta)$$
(14)

$$V_3(x_k) = \sum_{\theta = k - r_M}^{k - 1} \alpha^{-2\theta} \cdot x^T(\theta) Tx(\theta)$$
(15)

$$V_4(x_k) = \sum_{\theta=k-r(k)}^{k-1} \alpha^{-2\theta} \cdot x^T(\theta) Qx(\theta) + \sum_{l=-r_M+1}^{r_m} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot x^T(\theta) Qx(\theta)$$
(16)

$$V_{5}(x_{k}) = \sum_{l=-r_{m}+1}^{0} \sum_{\theta=k-l+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{1} \eta(\theta)$$
 (17)

$$V_{6}(x_{k}) = \sum_{l=-r(k)+1}^{0} \sum_{\theta=k-l+l}^{k-l} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta) + \sum_{l=-r(k)+1}^{-r_{m}} \sum_{\theta=k+l}^{k-l} \alpha^{-2\theta} \cdot (\theta-k-l+1) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$
(18)

$$V_{7}(x_{k}) = \sum_{l=-r_{k}+1}^{0} \sum_{\theta=k-1+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{3} \eta(\theta)$$

$$(19)$$

$$V_{8}(x_{k}) = \sum_{l=-r_{k}}^{-r_{m}-1} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{4} \eta(\theta)$$
(20)

where P>0, S>0, Q>0, T>0,  $R_i>0$ , i=1,2,3,4, and  $\eta(\theta)=x(\theta+1)-x(\theta)$ . Taking the difference of Lyapunov functionals  $V_j(x_k)$ , j=1,2,3,4,5,6,7,8, along the solutions of system (1) has the form

$$\Delta V_1(x_k) = \alpha^{-2k} \cdot \left[ \alpha^{-2} \cdot x^T(k+1) P x(k+1) - x^T(k) P x(k) \right]$$
(21)

$$\Delta V_2(x_k) = \alpha^{-2k} \cdot \left[ x^T(k) Sx(k) - \alpha^{2r_m} \cdot x^T(k - r_m) Sx(k - r_m) \right]$$

(22)

$$\Delta V_3(x_k) = \alpha^{-2k} \cdot \left[ x^T(k) T x(k) - \alpha^{2r_M} \cdot x^T(k - r_M) T x(k - r_M) \right]$$

$$\Delta V_4(x_k) = \sum_{\theta=k-r(k+1)+1}^k \alpha^{-2\theta} \cdot x^T(\theta) Q x(\theta)$$

$$- \sum_{\theta=k-r(k)}^{k-1} \alpha^{-2\theta} \cdot x^T(\theta) Q x(\theta)$$

$$+ \alpha^{-2k} \cdot (r_M - r_m) \cdot x^T(k) Q x(k)$$

$$- \sum_{\theta=k-r_M+1}^{k-r_m} \alpha^{-2\theta} \cdot x^T(\theta) Q x(\theta)$$
(24)

By some simple derivations, we have 
$$\sum_{\theta=k-r(k+1)+1}^{k} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta) - \sum_{\theta=k-r(k)}^{k-1} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$= \alpha^{-2k} \cdot x^{T}(k)Qx(k) + \sum_{\theta=k-r(k+1)+1}^{k-1} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$- \sum_{\theta=k-r(k)+1}^{k-1} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$- \alpha^{-2(k-r(k))} \cdot x^{T}(k-r(k))Qx(k-r(k))$$

$$\leq \alpha^{-2k} \cdot x^{T}(k)Qx(k) + \sum_{\theta=k-r_M+1}^{k-1} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$- \sum_{\theta=k-r_M+1}^{k-1} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$- \alpha^{-2(k-r(k))} \cdot x^{T}(k-r(k))Qx(k-r(k))$$

$$\leq \alpha^{-2k} \cdot x^{T}(k)Qx(k) + \sum_{\theta=k-r_M+1}^{k-r_M} \alpha^{-2\theta} \cdot x^{T}(\theta)Qx(\theta)$$

$$- \alpha^{-2(k-r(k))} \cdot x^{T}(k-r(k))Qx(k-r(k))$$

$$\Delta V_{5}(x_{k}) = \alpha^{-2k} \cdot r_{m} \cdot \eta^{T}(k)R_{1}\eta(k)$$

$$- \sum_{\theta=k-r_{m}}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta)R_{1}\eta(\theta)$$

$$- \sum_{\theta=k-r_{m}}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

$$- \sum_{\theta=k-r_{m}}^{k-1} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

$$+ \sum_{\theta=k+l+1}^{r_m} \sum_{\theta=k+l+1}^{k} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

$$- \sum_{\theta=k-r_M+1}^{k-1} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l+1) \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

$$- \sum_{\theta=k-r_M+1}^{k-1} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l+1) \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

$$- \sum_{\theta=k-r_M+1}^{k-1} \sum_{\theta=k-l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l+1) \cdot \eta^{T}(\theta)R_{2}\eta(\theta)$$

By some simple derivations, we have

$$\sum_{l=-r(k)+1}^{0} \sum_{\theta=k+l}^{k} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r(k)+1}^{0} \sum_{\theta=k-1+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$= \alpha^{-2k} \cdot r(k+1) \cdot \eta^{T}(k) R_{2} \eta(k)$$

$$+\sum_{l=-r(k)+1}^{0} \alpha^{-2\theta} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r(k)+1}^{0} \alpha^{-2(k-1+l)} \cdot \eta^{T}(k-1+l) R_{2} \eta(k-1+l)$$

$$-\sum_{l=-r(k)+1}^{0} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$\leq \alpha^{-2k} \cdot r_{M} \cdot \eta^{T}(k) R_{2} \eta(k) + \sum_{l=-r_{M}+1}^{0} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{k=k-r(k)}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta) - \sum_{l=-r_{M}+1}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$\leq \alpha^{-2k} \cdot r_{M} \cdot \eta^{T}(k) R_{2} \eta(k) - \sum_{\theta=k-r(k)}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$+\sum_{l=-r_{M}+1}^{-r_{M}} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$(26) \qquad =\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot (\theta-k-l) \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=-r_{M}+1}^{r_{M}} \sum_{\theta=k+l+1}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{2} \eta(\theta)$$

$$-\sum_{l=r, t=1}^{r_m} \sum_{\theta=k+l}^{k-1} \alpha^{-2\theta} \cdot \eta^T(\theta) R_2 \eta(\theta)$$
 (29)

$$\Delta V_{7}(x_{k}) = \alpha^{-2k} \cdot r_{M} \cdot \eta^{T}(k) R_{3} \eta(k)$$

$$- \sum_{\theta=k-r_{M}}^{k-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{3} \eta(\theta)$$
(30)

$$\Delta V_{8}(x_{k}) = \alpha^{-2k} \cdot (r_{M} - r_{m}) \cdot \eta^{T}(k) R_{4} \eta(k)$$

$$- \sum_{\theta=k-r_{M}}^{k-r_{K}-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{4} \eta(\theta)$$

$$- \sum_{\theta=k-r_{K}}^{k-r_{m}-1} \alpha^{-2\theta} \cdot \eta^{T}(\theta) R_{4} \eta(\theta)$$
(31)

Then we can obtain the following result from (21)-(31)

$$\Delta V(x_{k}) = \sum_{i=1}^{8} \Delta V_{i}(x_{k})$$

$$\leq \alpha^{-2k} \cdot \left\{ \alpha^{-2} \cdot x^{T} (k+1) P x (k+1) + x^{T} (k) \left[ -P + S + (r_{Mm} + 1) \cdot Q + T \right] x (k) \right\}$$

$$- \alpha^{2r_{m}} \cdot x^{T} (k - r_{m}) S x (k - r_{m})$$

$$- \alpha^{2r_{M}} \cdot x^{T} (k - r_{M}) T x (k - r_{M})$$

$$- \alpha^{2r_{M}} \cdot x^{T} (k - r (k)) Q x (k - r (k))$$

$$+ \eta^{T}(k) \left[ r_{m} \cdot R_{1} + \hat{r}_{Mm} \cdot R_{2} + r_{M} \cdot R_{3} + r_{Mm} \cdot R_{4} \right] \eta(k)$$

$$- \alpha^{2r_{m}} \cdot \sum_{\theta = k - r_{m}}^{k - 1} \eta^{T}(\theta) R_{1} \eta(\theta)$$

$$- \alpha^{2r_{M}} \cdot \left[ \sum_{\theta = k - r_{M}}^{k - 1} \eta^{T}(\theta) R_{2} \eta(\theta) + \sum_{\theta = k - r_{M}}^{k - 1} \eta^{T}(\theta) R_{3} \eta(\theta) \right]$$

$$+ \sum_{\theta = k - r_{M}}^{k - r(k) - 1} (\theta) R_{4} \eta(\theta) + \sum_{\theta = k - r(k)}^{k - r_{M} - 1} \eta^{T}(\theta) R_{4} \eta(\theta)$$
(32)

Define

$$\beta^{T}(k) = [x^{T}(k) \quad x^{T}(k-r_{m}) \quad x^{T}(k-r(k)) \quad x^{T}(k-r_{M})]$$

From LMIs in (9) with  $\sum_{\theta=k-r(k)}^{k-1} \eta(\theta) = x(k) - x(k-r(k))$ 

and the similar equations, we have

$$\sum_{\theta=k-r_{m}}^{k-1} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}^{T} \begin{bmatrix} R_{111} & R_{112} \\ * & R_{1} \end{bmatrix} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}$$

$$= r_{m} \cdot \beta(k)^{T} R_{111} \beta(k) + 2\beta(k)^{T} (k) R_{112} \sum_{\theta=k-r_{m}}^{k-1} \eta(\theta)$$

$$+ \sum_{\theta=k-r_{m}}^{k-1} \eta^{T}(\theta) R_{1} \eta(\theta) \ge 0$$
(33)

$$\sum_{\theta=k-r(k)}^{k-1} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}^{T} \begin{bmatrix} R_{211} & R_{212} \\ * & R_{2} \end{bmatrix} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}$$

$$= r(k) \cdot \beta^{T}(k) R_{211} \beta(k) + 2 \beta^{T}(k) R_{212} \sum_{\theta=k-r(k)}^{k-1} \eta(\theta)$$

$$+ \sum_{\theta=k-r(k)}^{k-1} \eta^{T}(\theta) R_{2} \eta(\theta) \ge 0$$
(34)

$$\sum_{\theta=k-r_{M}}^{k-1} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}^{T} \begin{bmatrix} R_{311} & R_{312} \\ * & R_{3} \end{bmatrix} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}$$

$$= r_{M} \cdot \beta^{T}(k) R_{311} \beta(k) + 2\beta^{T}(k) R_{312} \sum_{\theta=k-r_{M}}^{k-1} \eta(\theta)$$

$$+ \sum_{\theta=k-r_{M}}^{k-1} \eta^{T}(\theta) R_{3} \eta(\theta) \ge 0$$
(35)

$$\sum_{\theta=k-r_{M}}^{k-r(k)-1} \left[ \beta(k) \right]^{T} \left[ R_{411} \quad R_{412} \right] \left[ \beta(k) \right] \\
= \left( r_{M} - r(k) \right) \cdot \beta^{T}(k) R_{411} \beta(k) + 2 \beta^{T}(k) R_{412} \sum_{\theta=k-r_{M}}^{k-r(k)-1} \eta(\theta) \\
+ \sum_{\theta=k-r_{M}}^{k-r(k)-1} \eta^{T}(\theta) R_{4} \eta(\theta) \ge 0$$
(36)

$$\sum_{\theta=k-r(k)}^{k-r_{m}-1} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}^{T} \begin{bmatrix} R_{511} & R_{512} \\ * & R_{4} \end{bmatrix} \begin{bmatrix} \beta(k) \\ \eta(\theta) \end{bmatrix}$$

$$= (r(k) - r_{m}) \cdot \beta^{T}(k) R_{511} \beta(k) + 2\beta^{T}(k) R_{512} \sum_{\theta=k-r(k)}^{k-r_{m}-1} \eta(\theta)$$

$$+ \sum_{\theta=k-r(k)}^{k-r_{m}-1} \eta^{T}(\theta) R_{4} \eta(\theta) \ge 0$$
(37)

Assume the particular case  $\sigma(k,x(k))=q$ , then we can obtain the following result

$$\begin{split} &\Delta V(x_{k}) + \alpha^{-2k} \cdot \left\{ \alpha^{2r_{m}} \cdot \left[ r_{m} \cdot \beta^{T}(k) R_{111} \beta(k) \right. \right. \\ &+ 2\beta^{T}(k) R_{112}[x(k) - x(k - r_{m})] + \sum_{\theta = k - r_{m}}^{k - 1} \eta^{T}(\theta) R_{1} \eta(\theta) \right] \right\} \\ &+ \alpha^{-2k} \cdot \left\{ \alpha^{2r_{M}} \cdot \left[ r(k) \cdot \beta^{T}(k) R_{211} \beta(k) \right. \right. \\ &+ 2\beta^{T}(k) R_{212}[x(k) - x(k - r(k))] + \sum_{\theta = k - r(k)}^{k - 1} \eta^{T}(\theta) R_{2} \eta(\theta) \right] \right\} \\ &+ \alpha^{-2k} \cdot \left\{ \alpha^{2r_{M}} \cdot \left[ r_{M} \cdot \beta^{T}(k) R_{311} \beta(k) \right. \right. \\ &+ 2\beta^{T}(k) R_{312}[x(k) - x(k - r_{M})] + \sum_{\theta = k - r_{M}}^{k - 1} \eta^{T}(\theta) R_{3} \eta(\theta) \right] \right\} \\ &+ \alpha^{-2k} \cdot \left\{ \alpha^{2r_{M}} \cdot \left[ (r_{M} - r(k)) \cdot \beta^{T}(k) R_{411} \beta(k) \right. \right. \\ &+ 2\beta^{T}(k) R_{412}[x(k - r(k)) - x(k - r_{M})] + \sum_{\theta = k - r_{M}}^{k - r(k)} \eta^{T}(\theta) R_{4} \eta(\theta) \right] \right\} \\ &+ \alpha^{-2k} \cdot \left\{ \alpha^{2r_{M}} \cdot \left[ (r(k) - r_{m}) \cdot \beta^{T}(k) R_{511} \beta(k) \right. \right. \\ &+ 2\beta^{T}(k) R_{512}[x(k - r_{m}) - x(k - r(k))] + \sum_{\theta = k - r(k)}^{k - r_{H}} \eta^{T}(\theta) R_{4} \eta(\theta) \right] \right\} \\ &\leq \alpha^{-2k} \cdot \left\{ \alpha^{-2} \cdot x^{T}(k + 1) Px(k + 1) \right. \\ &+ x^{T}(k) \left( - P + S + (r_{M}) Tx(k - r_{M}) \right. \\ &- \alpha^{2r_{M}} \cdot x^{T}(k - r_{m}) Sx(k - r_{m}) \\ &- \alpha^{2r_{M}} \cdot x^{T}(k - r_{m}) Sx(k - r_{m}) \\ &- \alpha^{2r_{M}} \cdot x^{T}(k - r_{M}) Tx(k - r_{M}) \\ &- \alpha^{2r_{M}} \cdot x^{T}(k - r_{M}) Qx(k - r(k)) \\ &+ \eta^{T}(k) [r_{m} \cdot R_{1} + \hat{r}_{Mm} \cdot R_{2} + r_{M} \cdot R_{3} + r_{Mm} \cdot R_{4}] \eta(k) \\ &+ 2\beta^{T}(k) R_{112}[x(k) - x(k - r_{m})] \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{r(k) - r_{m} + r_{m}}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{211} \beta(k) \right. \\ &+ 2\beta^{T}(k) R_{312}[x(k) - x(k - r_{M})] \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{(r_{M} - r(k)) - r_{m} + r_{m}}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{411} \beta(k) \right. \\ &+ 2\beta^{T}(k) R_{412}[x(k - r(k)) - x(k - r_{M})] \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{(r(k) - r_{m})}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{411} \beta(k) \right. \\ &+ 2\beta^{T}(k) R_{412}[x(k - r(k)) - x(k - r_{M})] \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{(r(k) - r_{m})}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{511} \beta(k) \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{(r(k) - r_{m})}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{511} \beta(k) \right] \\ &+ \alpha^{2r_{M}} \cdot \left[ \left[ \frac{(r(k) - r_{m})}{r_{Mm}} \right] \cdot r_{Mm} \cdot \beta^{T}(k) R_{511} \beta(k) \right] \\ &+ \alpha^{2r_{$$

$$+2\beta^{T}(k)R_{512}[x(k-r_{m})-x(k-r(k))]\}$$

$$+2\beta^{T}(k)R_{512}[x(k-r_{m})-x(k-r(k))]\}$$

$$=\alpha^{-2k}\cdot[x^{T}(k)(\alpha^{-2}\cdot A_{q}^{T}PA_{q}-P+U)x(k)$$

$$+\beta^{T}(k)\cdot(\varepsilon(k)\cdot\overline{\Xi}_{1q}+(1-\varepsilon(k))\cdot\overline{\Xi}_{2q})\beta(k)]$$
(38)

Where  $\varepsilon(k) = \frac{r(k) - r_m}{r_{Mm}}$ 

$$\overline{\Xi}_{iq} = \Xi_{11q}^{i} + \begin{bmatrix} \overline{A}_{q}^{T} P \\ 0 \\ \overline{A}_{dq}^{T} P \\ 0 \end{bmatrix} (\alpha^{2} \cdot P)^{-1} \begin{bmatrix} \overline{A}_{q}^{T} P \\ 0 \\ \overline{A}_{dq}^{T} P \\ 0 \end{bmatrix}^{T} + \begin{bmatrix} (\overline{A}_{q} - I)^{T} \\ 0 \\ \overline{A}_{lq}^{T} \\ 0 \end{bmatrix} (r_{m} \cdot R_{1} + \hat{r}_{Mm} \cdot R_{2} + r_{M} \cdot R_{3} + r_{Mm} \cdot R_{4} \begin{bmatrix} (\overline{A}_{q} - I)^{T} \\ 0 \\ \overline{A}_{lq}^{T} \\ 0 \end{bmatrix}^{T}$$
(39)

with  $\Xi_{11q}^i$ ,  $W_j$ , i = 1,2, j = 1,2,3,4,5,  $q = 1,2,\cdots,N$ , are defined in (10).

Since  $0 \le \mathcal{E}(k) \le 1$ , the  $\mathcal{E}(k) \cdot \overline{\Xi}_{1q} + (1 - \mathcal{E}(k)) \cdot \overline{\Xi}_{2q}$  is a convex combination of  $\overline{\Xi}_{1q}$  and  $\overline{\Xi}_{2q}$ . Hence  $\mathcal{E}(k) \cdot \overline{\Xi}_{1q} + (1 - \mathcal{E}(k)) \cdot \overline{\Xi}_{2q} < 0$  will imply  $\overline{\Xi}_{1q} < 0$  and  $\overline{\Xi}_{2q} < 0$ 

Define

$$\hat{\Xi}_{iq} = \begin{bmatrix} \Xi_{11q}^i & \Xi_{12q} \\ * & \Xi_{22q} \end{bmatrix} + \Lambda_q F_q(k) \Gamma_q^T + \Gamma_q F_q^T(k) \Lambda_q^T$$
 (40)

where

$$\begin{split} & \Lambda_q^{\!{}^T} \! = \! \begin{bmatrix} 0 & 0 & 0 & M_q^{\!{}^T}P & M_q^{\!{}^T}R_1 & M_q^{\!{}^T}R_2 & M_q^{\!{}^T}R_3 & M_q^{\!{}^T}R_4 \end{bmatrix} \\ & \Gamma_q^{\!{}^T} \! = \! \begin{bmatrix} N_{1q} & 0 & N_{2q} & 0 & 0 & 0 & 0 & 0 \end{bmatrix} \!, \text{ and } \Xi_{11q}^i \,, \\ & \Xi_{12q} \,, \ \Xi_{22q} \text{ are defined in (10)}. \end{split}$$

By the condition (11) with Lemma 3 and the switching signal in (8), we can obtain the following result

$$x^{T}(k)(\alpha^{-2} \cdot A_{a}^{T}PA_{a} - P + U)x(k) < 0, \forall x(k) \in \overline{\Omega}_{a}$$
 (41)

By Lemmas 1-3 with above conditions, the conditions  $\Xi_{iq} < 0$ , i = 1,2, in (10) will imply  $\hat{\Xi}_{iq} < 0$ , i = 1,2, in (40), the conditions  $\hat{\Xi}_{iq} < 0$ , i = 1,2, in (40) will also imply  $\overline{\Xi}_{iq} < 0$  in (39).Therefore, the conditions  $\overline{\Xi}_{iq} < 0$ , i = 1,2, in (39) are equivalent to  $\varepsilon(k) \cdot \overline{\Xi}_{1q} + (1 - \varepsilon(k)) \cdot \overline{\Xi}_{2q} < 0$  in (38). From the condition (41) and the matrix  $\varepsilon(k) \cdot \overline{\Xi}_{1q} + (1 - \varepsilon(k)) \cdot \overline{\Xi}_{2q} < 0$  in (28) with (9), we have

$$\Delta V(x_k) = V(x_{k+1}) - V(x_k) \le 0$$
,  $k = 0, 1, 2, 3, \dots$   
 $V(x_{k+1}) \le V(x_k)$ ,  $k = 0, 1, 2, 3, \dots$ 

This implies

$$V(x_k) \le V(x_0), \ k = 0, 1, 2, 3, \dots,$$
  
 $\alpha^{-2k} \cdot \lambda_{\min}(P) \cdot ||x(k)||^2 \le V(x_k) \le V(x_0) \le \delta_1 \cdot ||x_0||^2,$ 

where

$$\begin{split} & \delta_{1} = \lambda_{\max}(P) + r_{m} \cdot \lambda_{\max}(S) + r_{M} \cdot \lambda_{\max}(T) \\ & + \left(r_{M}^{2} - r_{M}r_{m} + r_{m}\right) \cdot \lambda_{\max}(Q) + r_{m}^{2} \cdot \lambda_{\max}(R_{1}) \\ & + \left(r_{M}^{2} + \frac{1}{2}(r_{M} + r_{m} + 1) \cdot (r_{M} - r_{m}) \cdot (r_{M} - 1)\right) \\ & \cdot \lambda_{\max}(R_{2}) + r_{M}^{2} \cdot \lambda_{\max}(R_{3}) + \left(r_{M} - r_{m}\right) \cdot r_{M} \cdot \lambda_{\max}(R_{4}) \end{split}$$

By some simple derivations, we have

$$\|\mathbf{x}(k)\| \le \sqrt{\delta_1/\lambda_{\min}(P)} \cdot \alpha^k \cdot \|\mathbf{x}_0\|_{s}, \ k = 0, 1, 2, 3, \dots$$

Therefore, we conclude that the system (1) is exponentially stable with convergence rate  $0 < \alpha < 1$ . This completes this proof.

**Remark 2.** By setting  $\alpha = 1$  in Theorem 1, the global asymptotic stability of system (1) can be achieved.

**Remark 3.** In Theorem 1, some additional nonnegative inequalities in (33) - (37) are proposed.  $R_i > 0$ , j = 1,2,3,4,  $R_{i11} \in \Re^{4n \times 4n}$ ,  $R_{i12} \in \Re^{4n \times n}$ ,

i = 1,2,3,4,5, in (9), are determined appropriately by LMI toolbox of MATLAB.

## 4. Illustrative examples

**Example 1**. Consider the system (1) with no uncertainties and the following parameters: (Example 3.1 of [7])

$$N=2, \ A_{1} = \begin{bmatrix} 0.54 & 1.02 \\ -0.17 & -0.31 \end{bmatrix}, \ A_{d1} = \begin{bmatrix} -0.01 & -0.06 \\ 0.01 & 0.04 \end{bmatrix},$$

$$A_{2} = \begin{bmatrix} -0.01 & -0.06 \\ 0.01 & 0.04 \end{bmatrix}, \ A_{d2} = \begin{bmatrix} 0.11 & 0.18 \\ -0.03 & -0.04 \end{bmatrix}.$$
 (42)

In order to show the obtained results, the allowable delay upper bound and switching sets that guarantees the global exponential stability for system (1) with (42) is provided in Table 1.

The delay upper bound and switching sets of switched system (1) with (42)		
[7]	$\alpha = 1$ , $r_m = 1$ , $r_M = 5$ globally	
	asymptotically stable (choose $\alpha_1 = \alpha_2 = 0.5$ )	
	$\overline{\Omega}_{1} = \begin{cases} \left[ x_{1} & x_{2} \right]^{T} \in \Re^{2} : -12.0213x_{1}^{2} \\ -38.8254x_{1}x_{2} - 31.6356x^{2} < 0 \end{cases}$	
	$\left[ -38.8254x_1x_2 - 31.6356x^2 < 0 \right]$	
	, $\overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1$ .	
	$\alpha=1$ , $r_{\scriptscriptstyle m}=1$ , $r_{\scriptscriptstyle M}=187$ globally	
Our results	asymptotically stable ( $\alpha = 01$ and $\alpha = 09$ )	
	$\overline{\Omega}_{1} = \begin{cases} [x_{1}  x_{2}]^{T} \in \Re^{2} : 2.9225x_{1}^{2} \\ +10.1092x_{1}x_{2} + 7.7291x^{2} < 0 \end{cases}$	
	$, \overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1.$	
	$lpha=0.85$ , $r_{_{\! m}}=1$ , $r_{_{\! M}}=7$ globally	
	exponentially stable ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$\overline{\Omega}_{1} = \begin{cases} [x_{1}  x_{2}]^{T} \in \Re^{2} : 1.5344x_{1}^{2} \\ +5.7566x_{1}x_{2} + 5.0051x^{2} < 0 \end{cases}$	
	$\left[ +5.7566x_1x_2 + 5.0051x^2 < 0 \right]$	
	, $\overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1$ .	

Table 1. Some obtained results for our proposed results in this paper.

**Example 2.** Consider the system (1) with the following parameters: (Example 3.1 of [10])

$$N = 2, A_1 = \begin{bmatrix} 0.8 & 0.2 \\ 0 & 0.91 \end{bmatrix}, A_{d1} = \begin{bmatrix} -0.1 & 0 \\ -0.1 & -0.1 \end{bmatrix},$$

$$A_2 = \begin{bmatrix} 0.3 & 0 \\ 0 & 0.58 \end{bmatrix}, A_{d2} = \begin{bmatrix} 0.12 & 0 \\ 0.11 & 0.11 \end{bmatrix}, M_1 = M_2 = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix},$$

$$N_{A1} = N_{A2} = N_{B1} = N_{B2} = \begin{bmatrix} 0.01 & 0 \\ 0 & 0.01 \end{bmatrix}$$

$$(43)$$

In order to show the obtained results, the allowable delay upper bound that guarantees the global asymptotic stability for system (1) with (43) is provided in Table 2.

The delay upper bound and switching sets of switched system (1) with (43)		
[10]	$r_m=1, r_M=5$ (under arbitrary switching signal)	
[4]	$r_m=1, r_M=6$ (under arbitrary switching signal)	
Our result	$r_m = 1$ , $r_M = 11$ ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$\overline{\Omega}_{1} = \begin{cases} \left[ x_{1} & x_{2} \right]^{T} \in \Re^{2} : 18.6759x_{1}^{2} \\ + 0.6803x_{1}x_{2} + 0.4395x^{2} < 0 \end{cases}$ $, \overline{\Omega}_{2} = \Re^{2} - \overline{\Omega}_{1}.$	

Table 2. Some delay bounds to guarantee the global asymptotic stability ( $\alpha = 1$ ) of system.

**Example 3.** Consider the system (1) with the following parameters:

$$N = 2 \quad A_{1} = \begin{bmatrix} 0.7 & 0.2 \\ 0.1 & 0.71 \end{bmatrix}, \quad A_{d1} = \begin{bmatrix} -0.2 & -0.2 \\ -0.1 & -0.2 \end{bmatrix}$$

$$A_{2} = \begin{bmatrix} 0.4 & -0.01 \\ -0.01 & 0.58 \end{bmatrix}, \quad A_{d2} = \begin{bmatrix} 0.2 & 0.1 \\ 0.1 & 0.1 \end{bmatrix}$$

$$M_{1} = M_{2} = \begin{bmatrix} 0.12 & 0 \\ 0 & 0.12 \end{bmatrix}, \quad N_{A1} = N_{A2} = N_{B1} = N_{B2} = \begin{bmatrix} 0.01 & 0 \\ 0 & 0.01 \end{bmatrix}$$

$$(44)$$

In order to show the obtained results, the allowable delay bounds that guarantee the global asymptotic stability for system (1) with (44) is provided in Table 3.

The delay upper bound and switching sets of switched system (1) with (44)		
[10]	$r_m = 1$ , $r_M = 2$	
	(under arbitrary switching signal)	
[4]	$r_m = 1$ , $r_M = 3$	
	(under arbitrary switching signal)	
	$r_m = 2$ , $r_M = 3$	
	(under arbitrary switching signal)	
	$r_m = 1$ , $r_M = 4$ (	
	$\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$\overline{\Omega}_1 = \begin{cases} [x_1  x_2]^T \in \Re^2 : 3.4451x_1^2 \\ +3.4325x_1x_2 + 1.7489x^2 < 0 \end{cases}$	
	$\left[ +3.4325x_1x_2 + 1.7489x^2 < 0 \right]$	
Our	, $\overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1$ .	
result	$r_m=2$ , $r_M=4$ (	
	$\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$\overline{\Omega}_{1} = \begin{cases} \left[ x_{1} & x_{2} \right]^{T} \in \Re^{2} : 1.2306x_{1}^{2} \\ +1.5704x_{1}x_{2} + 0.7528x^{2} < 0 \end{cases}$	
	$\left[ +1.5704x_1x_2 + 0.7528x^2 < 0 \right]$	
	, $\overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1$ .	

Table 3. Some delay bounds to guarantee the global asymptotic stability ( $\alpha = 1$ ),of system.

**Example 4.** Consider the system (1) with the following parameters: (Example of [9])

$$N = 2 \quad A_{1} = \begin{bmatrix} 0.7 & 0 \\ 0.08 & 0.95 \end{bmatrix}, \quad A_{d1} = \begin{bmatrix} 0.15 & 0 \\ -0.1 & -0.1 \end{bmatrix},$$

$$A_{2} = \begin{bmatrix} 0.7 & 0 \\ 0.08 & 0.9 \end{bmatrix}, \quad A_{d2} = \begin{bmatrix} 0.14 & 0 \\ -0.1 & -0.1 \end{bmatrix}, \quad M_{1} = \begin{bmatrix} 0.05 \\ 0 \end{bmatrix},$$

$$M_{2} = \begin{bmatrix} 0.05 \\ -0.02 \end{bmatrix}, \quad N_{11} = \begin{bmatrix} 0.2 & 0.3 \end{bmatrix}, \quad N_{12} = \begin{bmatrix} -0.1 & -0.1 \end{bmatrix}$$

$$N_{21} = \begin{bmatrix} 0 & -0.1 \end{bmatrix}, \quad N_{22} = \begin{bmatrix} -0.3 & -0.2 \end{bmatrix}. \tag{45}$$

In order to show the obtained results, the allowable delay bounds that guarantee the global asymptotic stability for system (1) with (45) is provided in Table 4.

The delay upper bound and switching sets of switched system (1) with (45)		
[9]	$r_m = 2$ , $r_M = 6$	
	(under arbitrary switching signal)	
[4]	$r_m = 1, r_M = 10$	
	(under arbitrary switching signal)	
	$r_m = 2$ , $r_M = 10$	
	(under arbitrary switching signal)	
	$r_m = 4$ , $r_M = 11$	
	(under arbitrary switching signal)	
	$r_m = 8, r_M = 12$	
	(under arbitrary switching signal)	
	$r_m = 1$ , $r_M = 11$ ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$ ([x   x]^T \in \Re^2 : -0.7576x^2)$	
	$\overline{\Omega}_{1} = \begin{cases} [x_{1}  x_{2}]^{T} \in \Re^{2} : -0.7576x_{1}^{2} \\ +18.2930x_{1}x_{2} + 50.3305x^{2} < 0 \end{cases}$	
	$, \overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1 .$	
	$r_m = 2$ , $r_M = 11$ ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
Our result	$r_{m} = 2, r_{M} = 11 (\alpha_{1} = 0.1 \text{ and } \alpha_{2} = 0.9)$ $\overline{\Omega}_{1} = \begin{cases} [x_{1} & x_{2}]^{T} \in \Re^{2} : -0.1590x_{1}^{2} \\ +1.4113x_{1}x_{2} + 4.4661x^{2} < 0 \end{cases}$	
	$\left[ +1.4113x_1x_2 + 4.4661x^2 < 0 \right]$	
	$\Omega_{2}=\Re^{2}-\Omega_{1}$ .	
	$r_m = 4$ , $r_M = 13$ ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$r_{m} = 4, r_{M} = 13 (\alpha_{1} = 0.1 \text{ and } \alpha_{2} = 0.9)$ $\overline{\Omega}_{1} = \begin{cases} [x_{1}  x_{2}]^{T} \in \Re^{2} : -0.7476x_{1}^{2} \\ +9.8448x_{1}x_{2} + 29.7229x^{2} < 0 \end{cases}$	
	$+9.8448x_1x_2 + 29.7229x^2 < 0$	
	, $\overline{\Omega}_2 = \mathfrak{R}^2 - \overline{\Omega}_1$ .	
	$r_m = 8$ , $r_M = 16$ ( $\alpha_1 = 0.1$ and $\alpha_2 = 0.9$ )	
	$\overline{\Omega}_{1} = \begin{cases} [x_{1}  x_{2}]^{T} \in \Re^{2} : -0.4987x_{1}^{2} \\ +2.8498x_{1}x_{2} + 8.0945x^{2} < 0 \end{cases}$	
	$\left[ +2.8498x_1x_2 + 8.0945x^2 < 0 \right]$	
	, $\overline{\Omega}_2 = \Re^2 - \overline{\Omega}_1$ .	

Table 4. Some delay bounds to guarantee the global asymptotic stability (  $\alpha=1$  ) of system.

Note that comparing the proposed examples, we can design a new switching signal to guarantee the global exponential stability for system (1). At the same time, less conservative results are reached without tuning any parameters.

## 5. Conclusions

In this paper, the global exponential stability for uncertain discrete switched systems with interval time-varying delay has been considered. Some additional nonnegative inequalities and LMI approach are used to guarantee the proposed results of the system. The obtained results have shown to be less conservative via some numerical examples. In [19-20], the practical applications have been successfully designed in the different control fields.

### Acknowledgments

The research reported here was supported by the National Science Council of Taiwan, R.O.C. under grant no. NSC102-2221-E-507-006.

## References

- [1] Z. Sun and S. S. Ge, "Switched Linear Systems Control and Design," Springer-Verlag, London, 2005.
- [2] Y. Han and H. Tang, "Robust control for a class of discrete switched systems with uncertainties and delays," Proc. 26th Chinese Control Conference, Hunan, China, pp. 681-684, 2007.
- [3] S. Ibrir, "Stability and robust stabilization of discretetime switched systems with time-delays: LMI approach," Applied Mathematics and Computation, Vol. 206, pp. 570-578, 2008.
- [4] C. H. Lien et al., "Exponential stability and robust control for uncertain discrete switched systems with interval time-varying delay," IMA Journal of Mathematical Control and Information, Vol. 7, pp. 433-444, 2011.
- [5] C. H. Lien et al., "Switching signal design for global exponential stability of uncertain switched nonlinear systems with time-varying delay," Nonlinear Analysis: Hybrid Systems, Vol. 5, pp. 10-19, 2011.
- [6] J. Liu et al., "Delay-dependent robust control for uncertain switched systems with time-delay," Nonlinear Analysis: Hybrid Systems, Vol. 2, pp. 81-95, 2008.
- [7] V. N. Phat and K. Ratchagit., "Stability and stabilization of switched linear discrete-time systems with interval time-varying delay," Nonlinear Analysis: Hybrid Systems, Vol. 5, pp. 605-612, 2011.
- [8] X. M. Sun et al., "Stability analysis for linear switched systems with time-varying delay," IEEE Transactions on Systems, Man, and Cybernetics, Part B, Vol. 38, pp. 528–533, 2008.
- [9] L. Zhang et al., Robust Stability and Stabilisation of Uncertain Switched Linear Discrete Time-Delay Systems, IET Control Theory Applications, Vol. 2, pp. 606-614, 2008.
- [10] Y. G. Sun et al., Delay-dependent robust stability and control for uncertain discrete-time switched systems with mode-dependent time delays, Applied Mathematics and Computation , Vol. 187, pp. 1228-1237, 2007.
- [11] D. Xie et al., Stabilisability and observer-based switched control design for switched linear systems, IET Control Theory Applications, Vol. 2, pp. 192-199, 2008.
- [12] G. Zhai et al., "Lie algebraic stability analysis for switched systems with continuous-time and discrete-time subsystems," IEEE Trans. Circuits Systems. Vol. 53, pp. 152–156, 2006.

- [13] L. Zhang et al., "Robust stability and stabilisation of uncertain switched linear discrete time-delay systems," IET Control Theory Applications, Vol. 2, pp. 606-614, 2008.
- [14] W. A. Zhang and L. Yu., Stability analysis for discrete-time switched time-delay systems, Automatica, Vol. 45, pp. 2265-2271, 2009.
- [15] R. S. Gau et al., "Novel stability conditions for interval delayed neural networks with multiple time-varying delays," International Journal of Innovative Computing, Information and Control , Vol. 7, pp. 433-444, 2011.
- [16] T. Li et al., "Robust stability for neural networks with time-varying delays and linear fractional uncertainties," Neurocomputing, Vol. 71, pp. 421-427, 2007.
- [17] K. Gu et al., "Stability of Time-Delay Systems," Birkhauser, Boston, Massachusetts, 2003.
- [18] S. P. Boyd et al., "Linear Matrix Inequalities in System and Control Theory," SIAM, Philadelphia, 1994.
- [19] S. L. Chu, M. J. Lo., "A New Design Methodology for Composing Complex Digital Systems," Journal of Applied Research and Technology, vol. 11, pp. 195-205, 2013.
- [20] Y. L. Chen and Z. R. Chen., "A PID Positioning Controller with a Curve Fitting Model Based on RFID Technology," Journal of Applied Research and Technology, vol. 11, pp. 301-310, 2013.